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Distributed Smart Sensing Systems for Inc By Octavian Postolache, Jose Migue

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#### Introduction

Indoor air quality pollution [ ][2] represents one of the factors associated with the etiology of chronic obstructive pulmonary disease and also plays an important role in respiratory distress, the second most common symptom of adults that request emergency transportation to the hospital and is associated with a relatively high overall mortality before hospital discharge [ ][ ][ ][ ]. The prevention of acute respiratory distress or asthma attacks can be possible by monitoring the air quality conditions using distributed smart sensing systems characterized by accuracy, short time response, and robustness as well as by data processing, data logging and data communication capabilities.

Considering the importance of indoor air quality monitoring, different distributed measuring system architectures and associated calibration methods and systems are presented in the literature [ ][ ][ ]. The main elements of these kind of systems are not only temperature and relative humidity sensors, but also gas detectors and gas concentration sensors whose metrological characteristics, such as accuracy and linearity are very limited, which implies the design and implementation of signal processing algorithms namely for numerical linearization and common factors correction [ ][ ][ ].

Taking into account the indoor spatial distribution of the temperature and relative humidity values as well as the concentration values of pollutants (e.g CO, CO<sub>2</sub> resulting of combustion), the development of distributed measuring systems [ 1] 1 that can include perconal computers (PCc) or mobile devices (a.g. PDAs [ or smart phones [ ]) based human-sensing system interface represents

This chapter presents a practical approach concerning distributed smarr the authors in this area. The first part of the chapter deals with the relat conditions. The second part contains a brief presentation of solid state s the third part presents a distributed architecture based on an embedded fourth part, a Bluetooth wireless distributed system including smart sen monitoring system interfacing device is presented.

Referring to the distributed air quality monitoring system based on emb oxide semi-conductor sensor array that includes general air contaminan pollution alarms from the nodes, which are parts of a wired or wireless temperature and relative humidity sensors are included in the node's har and humidity influences. This chapter also includes a brief description  $\boldsymbol{\varepsilon}$ used to obtain temperature and humidity compensated gas concentratic designed and implemented for continuous monitoring of indoor humid the chapter. Bluetooth compatible nodes, characterized by data acquisiti using Java2ME to perform different tasks including data communication indoor air quality monitoring system. Elements regarding the smart pho architecture and air quality monitoring tasks are discussed and an exam system, an intelligent assessment of air conditions for risk factor reduct

2. Air quality and its impact on respiratory diseases

Air conditions and respiratory assessment represent an important challenge taking into account that distress is the second most common symptom of adults transported by ambulance and is associated with a relatively high overall mortality before hospital discharge []. Among the most common causes of respiratory distress in this setting are congestive heart failure, pneumonia, chronic obstructive pulmonary disease and asthma []. It is projected that chronic obstructive pulmonary disease (COPD) will be the third leading cause of death worldwide by 2020, due to an increase in smoking rates and demographic changes in many countries []. Worldwide, some 300 million people currently suffer from asthma. It is the most common chronic disease among children []. The economic burden of COPD in the US in 2007 was 42.6 billion in health care costs and lost productivity []. The indoor air pollution is one of the factors associated with specific "asthma genes" are crucial for the development of chronic, persistent form of disease [][]. The identification of the indoor air associated with pathophysiology of COPD and asthma disease will thus be crucial for the primary-prevention strategy.

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Greece	\$37,000	3.1		
Portugal	\$44,000	3.5	\$64,000	3.5
Czech Republic	\$35,000	1.7	\$32,000	1.7
Hungary	\$27,000	1.7	\$26,000	1.6
Moxico	925,000	2.4	\$21,000	21
Polood	\$20,000	1.6		
Average	\$113,000	3.7	883,000	2.9
Average excluding U.S.	\$107.000	3.6	\$79,000	2.8
Median	\$83,000	3.3	580,080	3.0

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3. Air quality sensing and data processing This section contains the description of the main components of a distri attention is dedicated to the implementation of the sensing nodes, to sig

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#### A. Sensing nodes

The sensing nodes are designed and implemented to perform the air quality (AirQ) monitoring using low cost gas sensors and, at the same time, to get additional information about the temperature (T) and relative humidity (RH). This information is used to increase gas concentration measurement accuracy, performing the error compensation caused by temperature and humidity influence.

The gas sensors can be sintered SnO<sub>2</sub> semiconductor heated sensors, as those provided by Figaro [ ], that assure pollution event detection (TGS800 – general air  $contaminant \ sensor - AC), \ methane \ detection \ (TGS842-M), \ alcohol \ and \ organic \ solvent \ detection \ (TGS822-SV) \ and \ carbon \ monoxide \ detection \ (TGS8203-CO).$ Information about temperature and relative humidity are obtained using Smartec SMT160-30 [ ] and Humirel HM1500 [ ] temperature and relative humidity transducers, respectively.

The gas sensors, connected to proper conditioning circuits, are devices that produce voltages whose values depend on the concentrations of gas expressed in ppm. The used conditioning circuit for the air pollution sensor TGS800, solvent vapors (TGS822) and methane sensor (TGS842) are presented in

Electrochemical cells can also be used to implement the sensing units. The NAP-505 [ ] is a typical example of this kind of implementation. In this cased, the 3 terminals measuring cell consists of 3 porous noble metal electrodes separated by an acidic aqueous electrolyte, housed within a plastic enclosure. The working principle of the sensing unit is based on chemical reactions between gas and other elements. From the electrical charges that are involved in those reactions it is possible to measure an electrical current that is proportional to gas concentration. Using multiple cells it is possible to measure the concentration of different gas

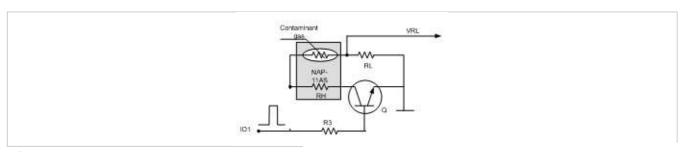


Figure 1. Gas sensing unit based on semiconductor hea VGS – gas sensor output voltage, RL – load re

represents the main elements of a gas sensing unit based on an

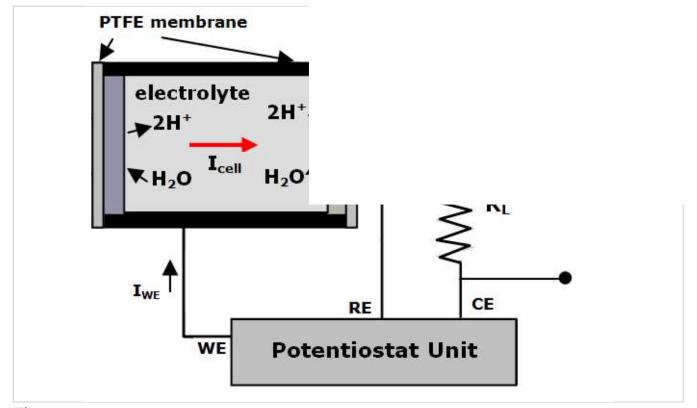


Figure 2.

Gas sensing unit based on an electrochemical cell (RE - reference electrode, CE - counting electrode, WE - working electrode, RL – load resistance)

The measuring cell includes a working electrode (WE), a counter electrode (CE) and a reference electrode (RE) [ ]. The conditioning circuit is basically a potentiostat unit that measures the gas dependent current amplitude ( $I_{cell}$ ) that flows between the CE and WE through cell's electrolyte. The current amplitude is directly proportional to the gas concentration but its value is usually very low, about a few tens of nA. For this reason a careful design of the potentiostat is crucial to obtain an acceptable measurement. represents the electrical diagram of a typical potentiostat conditioning circuit [ ][ ]. The negative feedback loop, provided by operational amplifiers (OA1 and OA2) and the electrical connection that exists between CE and RE electrodes through the sensing element, assures that the operational amplifiers are working in their linear zones. Since the current between the working and the reference electrodes is very low, the differential voltage between working and counter electrodes is equal to VRE and the output voltage (VADC) from the current to voltage converter implemented by sub-circuit 2 is given by

$$V_{ADC} = -R_F \cdot [f_{sol.}(V_{DAC}) - I_B]$$

where RF represents the feedback resistor of the current to voltage converter, IB represents the polarization current of OA2, VDAC is the output voltage of the D/A converter and fsol is generally a non-linear function that depends on solution characteristics and applied voltage (VWE).

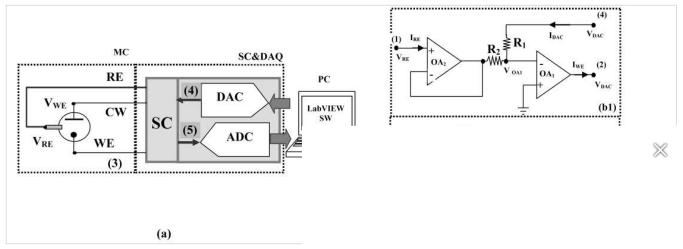


Figure 3.

Electrical circuit of a voltammetry measuring CE- counting electrode, WE- working electro ADC- analogue to digital converter, DAC- dig Another attractive solution that can be used to implement the sensing n interdigitated transducer etched onto a piezoelectric substrate, covered chemical substance from the air. This causes a shift in resonance to a sli

#### B. Measurement data interpolation

To perform the interpolation of the calibration data in order to obtain the namely, polynomial interpolation and artificial neural networks (ANNs

Assuming, for simplicity, a single variable function (f) and a LMS polyr

$$P_n(x) = \sum_{k=0} \alpha_k \cdot x^k$$

E2

where *p* represents the degree of the polynomial curve fitting function and *x* represents the independent variable - measured quantity - it is possible to demonstrate that the LMS deviation between calibration and curve fitting data is obtained when the coefficients of the curve fitting polynomial function are given by

$$[\alpha] = \left[\mathbf{X}_{\mathbf{C}}^{\mathbf{T}} \cdot \mathbf{X}_{\mathbf{C}}\right]^{-1} \cdot \left[\mathbf{X}_{\mathbf{C}}^{\mathbf{T}} \cdot \mathbf{Y}\right]$$

E

being vector Y and matrix X<sub>C</sub> defined, for a set or n calibration points, by

$$\mathbf{Y} = egin{bmatrix} \mathbf{y}_1 \\ \mathbf{y}_2 \\ \vdots \\ \mathbf{y}_n \end{bmatrix} \mathbf{X}_{\mathrm{C}} = egin{bmatrix} 1 \ x_1 \ x_1^2 \dots x_1^p \\ 1 \ x_2 \ x_2^2 \dots x_2^p \\ \vdots \ \vdots \ \vdots \ \dots \ \vdots \\ 1 \ x_n \ x_n^2 \dots x_n^p \end{bmatrix}$$

E4

The concerns related with polynomial interpolation are mainly associated with the choice of the polynomial degree. If a low polynomial degree is used, the interpolation error is generally high because the polynomial function can not fit correctly a large number of calibration points. Conversely, if an excessive polynomial degree is used, the LMS deviation between calibration data and the values obtained from the polynomial interpolation function may be very low, but the interpolation errors of points between calibration data are usually very high. This problem is usually known as overfitting and the previous one as underfitting.

Regarding ANN [ ][ ][ ], the curve fitting function can be computed using the following expression:

$$F_{ANN}(x_i) = F_N \left( W_N * \left( F_{N-1} \left( ... F_2 \left( W_2 * F_1 \! \left( W_1 * x_i + B_1 \right) + B_2 \right) ... + B_{N-1} \right) + B_N \right) \right.$$

E5

where N represents the number of neural network (NN) layers, Bi the bias vectors, Wi the weight vectors and Fi the activation transfer function of each layer.

The most common ANN structure for measurement applications contains a hidden layer of neurons with sigmoidal activation functions whose input is the measured data, and an output layer of neurons with linear activation functions. This ANN structure calculates an output vector given by

$$F_{ANN}(x_i) = purelin(W_2*tansig(W_1*x_i + B_1) + B_2)$$

E6

where purelin() and tansig() are linear and hyperbolic tangent sigmoidal activation transfer functions, respectively.

This architecture has proved capable of approximating any function with a finite number of discontinuities and with arbitrary accuracy. Generally a more complex function, such as transducer characteristics that are strongly non-linear, requires more sigmoidal neurons in the hidden layer.

To evaluate the capability of a given solution to generalize the learned fit correspondent interpolated errors are evaluated. The best values of [B] by minimizing the mean square error



MSE(x)

Ε,

Several gradient methods  $[\ ][\ ]$ , like back propagation (generalized  $\Delta$  set of input values corresponding to the calibration points is used to adjudy output and the calibration values.

Even if there is no general rule to choose polynomial or ANN based curvare available, and especially when extrapolation capabilities are desired.

particularly true for non-linear and non-deterministic sensors' characte terms of the computational load caused by an higher number of mathen (tanh()) [ ][ ][ ].

C. Data processing: an application example

In order to take advantage of the joint use of polynomial and artificial neural network (ANN) curve fitting techniques [ ][ ], this section describes a hybrid solution based on polynomial modelling (PM) and artificial neural networks modelling (ANN-M) that can be used to estimate the values of air quality parameters, such as, temperature, relative humidity, and polluting gases concentration.

For the particular case of broadband gas sensors, different methods can be used to convert the measured data into concentration of possible gas contaminants, such as, methane, carbon monoxide, isobutane, hydrogen, ethanol or cigarette smoke. Considering the voltage generated by a gas sensing unit based on a semiconductor heated sensor (TGS800 from Figaro), an air quality index  $\zeta$ , is defined using the following relation

$$\zeta = \frac{R_S}{R_{S0}} = \left(\frac{V_C}{V_{RL}} - 1\right) \cdot \frac{1}{\left(\frac{V_C}{V_{RL0}} - 1\right)}$$

where  $R_{SO}$  represents the sensor resistance for a clean air condition,  $R_{S}$  represents the sensor resistance for the tested air,  $V_{C}$  is the circuit power supply voltage,  $V_{RL}$  is the load resistor voltage and  $V_{RLo}$  is the load resistor voltage for clean air.

Sensor's characteristic is non-linear and monotonic, decreasing sensor's resistance ratio with contaminant gas concentration. Higher concentrations of contaminants originate lower values of resistance ratios. Moreover, since the sensor is designed for general contaminants detection, it is not possible to identify specific contaminants. So, according to the application requirements in terms of the maximum acceptable level of contamination, a coefficient (ζ) value equal to 0.3 is considered for air pollution alarm. Considering that the used sensor has not good selectivity for each potential air contaminant, a look-up table, a polynomial, and a multilayer perceptron single-input single output neural network were designed and implemented to convert the value of  $\zeta$  into air contaminants' concentrations expressed in parts per million (ppm). The measurement data processing scheme that was implemented is represented in

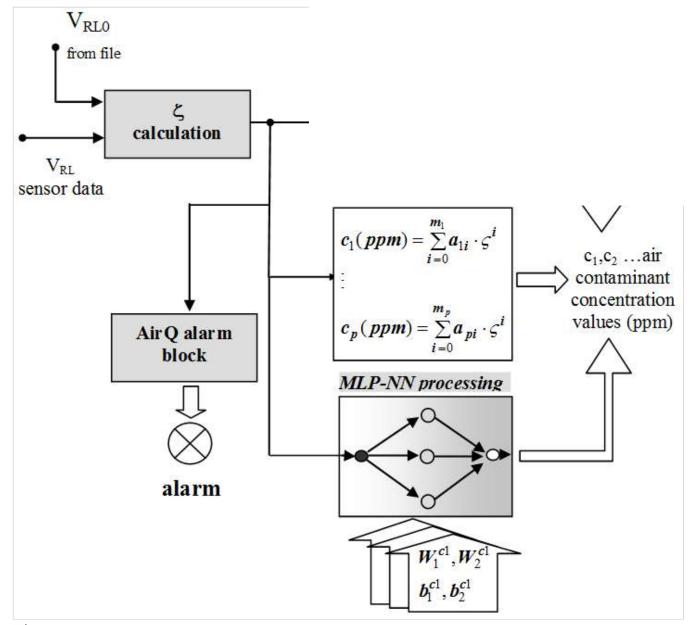


Figure 4. Block diagram of the hybrid data processing scheme that was used to evaluate contaminants' air concentrations.

To test the performance of the proposed modeling scheme, a set of coefficient values ( $\zeta$ ), contained in the interval between 0.15 and 1 (no pollution), and the correspondent values of air contaminants' concentrations obtained from TGS800 sensitivity curves for methane, carbon monoxide, isobutane, hydrogen and ethanol, were considered. The calculation of polynomial coefficients,  $a_{1i}$ ,  $a_{2i}$ , ...  $a_{pi}$ , is based on LS linear fit function (Givens method) that is implemented in LabVIEW. The calculated polynomial coefficients values that correspond to TGS800 sensitivity curves, such as the ones represented in memory and then used to perform the evaluation of air contaminants' concentrations.

The used neural processing blocks (NPB<sub>i</sub>) is related with the inverse modeling [ ] of gas sensor multivariable nonlinear characteristics, which are strongly dependent on temperature and humidity but also influenced by the concentration of other gases of the analyzed gas mixture. Based on the designed NPB<sub>i</sub>, a digital read-out of the gases concentration with temperature and compe

Regarding the NPB<sub>i</sub>, two inputs one output multilayer perceptron neura normalization blocks and denormalization blocks used for ANN input a

The NPBi's internal parameters (weights and biases) are off-line calculated system calibration phase. They are voltage values ( $V_{GSi}$ ) acquired from  $(C_{\mbox{\scriptsize Gi}}),$  and different temperature  $(T_{\mbox{\scriptsize p}})$  and relative humidities  $(RH_{\mbox{\scriptsize i}})$  val

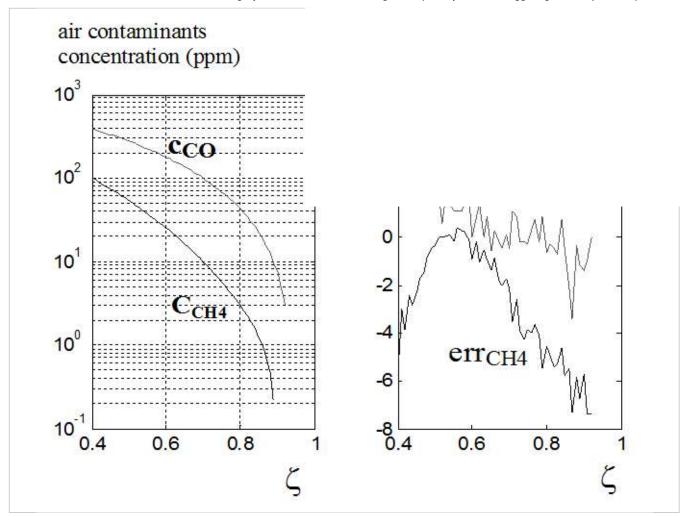


Figure 5. Polynomial approximation of air contaminants curves (CO and methane case) and polynomial approximation error (err<sub>CO</sub>, err<sub>CH4</sub>)

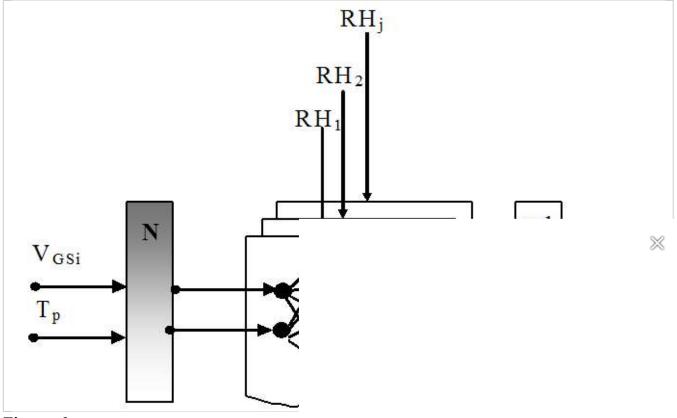


Figure 6.

NPB; architecture (N, N<sup>-1</sup>: normalization and temperature and humidity compensated valu  $V_{GS_i}$ : input voltage value on the  $GS_i$  channel)

The neural network algorithm developed in MATLAB software calculate 55%, 65%}. The NPB; input is the normalized voltage associated with ear temperature compensated gas concentration ( $C_{Gi}$ ). The NPBi normalize

$$V_{GSi}^{N}=\cdot$$

where  $V_{1S}$  represents the gas sensor normalization factor (GS<sub>i</sub> voltage supply=+10V in the present case).

Because GSi characteristics depend on humidity, an accurate measurement of the gas concentration is provided using different NPB  $_{ilRH}$  whose weights and biases are calculated using the data obtained for predefined relative humidity conditions (RH=45%, 55% and 65%) and by the interpolation method presented in [ \_\_].

The number of NPBi's layers is three. The hidden layers have 2 to 5 tansignoid (tansig(x)) neurons, and the output layer has 1 linear (l(x)) neuron. The implemented tansig(x) calculates its output according to

$$\tan sig(x) = \frac{2}{1 + \exp(-2x)} - 1$$

E10

which leads to a reduction of the computational load.

Two criteria for NPBi design were considered, the type and the number of neurons on the hidden layer, both determining the capabilities of the NPBi to adapt to a given characteristic. Different neuron nonlinear activation functions require different memory space and processing capabilities from the hardware platform.

To reduce the weights and biases in vector sizes, several simulation tests concerning the number of neurons for a required NPB; performance, expressed by a modeling error, were performed. ANNs with a higher number of neurons increase processing load and, moreover, require larger memories to store weights and biases matrices. The results of these simulations are particularly important when embedded systems are used to implement the neural processing architecture (e.g. 512k EEPROM in the IPµ8930 case).

For the particular case of the CO measuring channel, the training set includes, as target, fifteen CO concentration values uniformly distributed in the 30 to 300ppm interval. The input values are the voltage values acquired from the TGS203 CO concentration measuring channel corresponding to the above-mentioned concentrations. The measured temperature in the testing chamber was  $T_p[^{\circ}C] = 10 \times p$ ,  $p = \{1,2,3,4,5\}$  and the relative humidity RH=35%. The Levenberg Marquardt algorithm [ ] was used to calculate the weights and biases (W<sub>NPBi</sub>, B<sub>NPBi</sub>) of the neural network. Imposing a sum square error stop condition SSE=0.01, and for  $neural\ networks\ characterized\ by\ 4,5\ or\ 6\ hidden\ neurons,\ different\ measuring\ channel\ modeling\ error\ characteristics\ (e_{CGsi})\ were\ obtained\ (e_{CGsi})$ modeling error is defined by:

$$e_{CGsi} = rac{C_{CGsi} - C_{CGsi}^{NPB}}{FS} imes 100$$

where FS represents the measurement range, CCGsi is the experimental used gas concentration (e.g. carbon monoxide concentration) expressed in ppm, and  $C_{CGsi}^{NPB}$  the concentration of gas calculated by the corresponding neural processing module.

Since the used gas sensors characteristic depends on temperature, a study related with the CO channel modeling error (e<sub>CO</sub>) versus temperature was carried out



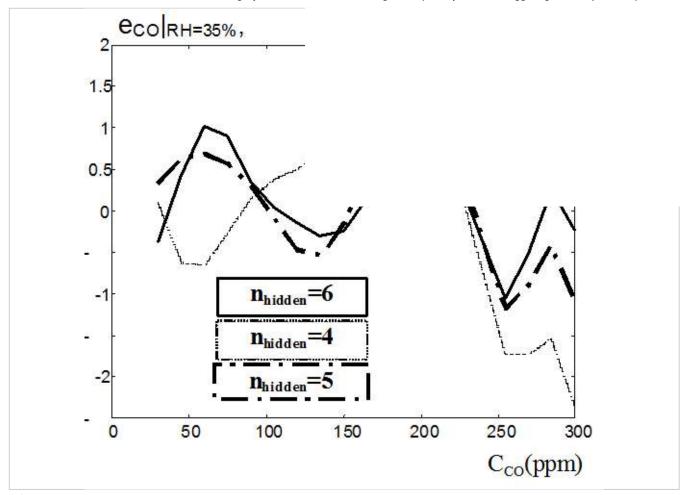


Figure 7. The modeling error versus concentration for different NPB $_{\rm CO}$  architectures (T=10°C)

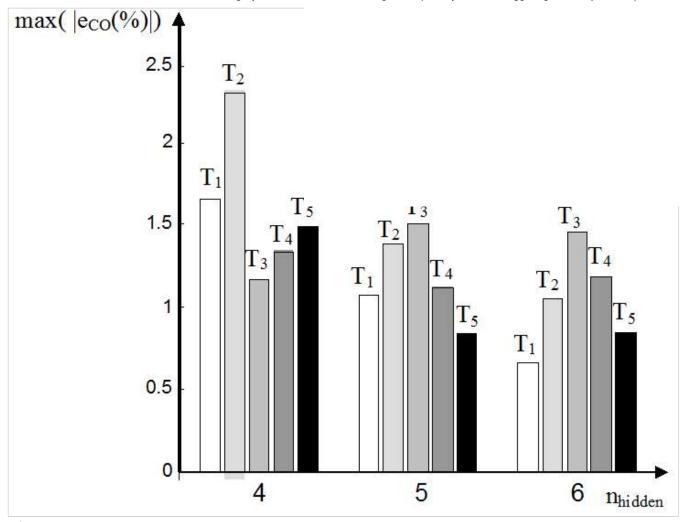


Figure 8. The maximum inverse modeling error for different NPB<sub>CO</sub> architectures ( $n_{hidden}$ ={4, 5, 6}) and different temperatures  $T_p=10 p °C$ 

Being humidity an influence quantity, different values of the relative humidity lead to different primary gas selectivity characteristics and hence to different gas concentration measurement accuracies. Thus, experimental data obtained for three different values of relative humidity, RH<sub>1</sub>=35%, RH<sub>2</sub>=65% and RH<sub>3</sub>=95%, and five values of temperatures included in the  $I_T$ =[10;50] $^{o}$ C were considered. The imposed gas concentrations for measurement system testing were: 10 values of methane concentration distributed in the  $I_{CM}$ =[500;5000] ppm interval, 15 values of carbon monoxide concentration  $I_{CCO}$ =[30;300] ppm, and 15 values of solvent vapors (Ethanol vapors) concentration,  $C_{SV}$ =[50;5000] ppm.

Based on the GS<sub>i</sub> voltages for the considered gases concentrations, and taking into account temperature and humidity, three sets of weights and biases (35%, 65% and 95% relative humidity) were calculated for carbon monoxide, methane and solvent vapor measurement channels.

architectures, including sensing nodes, materialize the implementations in the air quality monitoring for indoor and outdoor conditions.



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<sup>4.</sup> Smart sensing networks for air quality assessment Gas sensors networks provide a promising mechanism for mining information from the monitored areas. Point-to-point and multipoint wireless network



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A. Point-to-point network architecture Different architectures were developed by the authors, one of them base (AIR-Q VMS) that joins hardware and software components to assure h the mobile indoor air quality monitor system is presented in

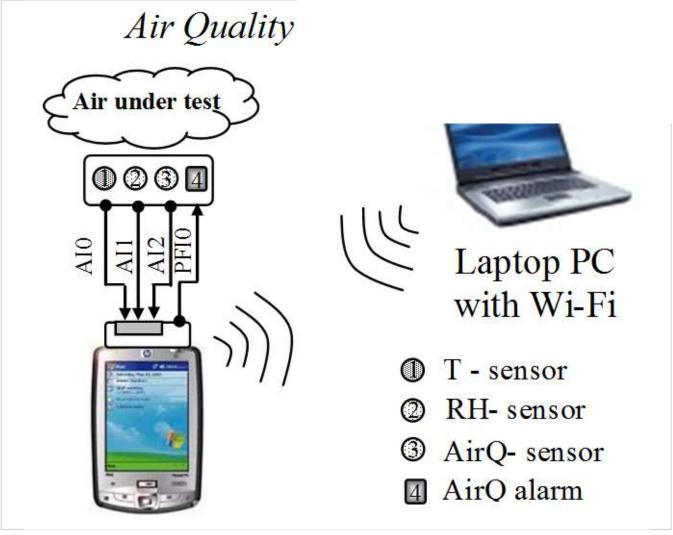


Figure 9. Mobile Air Quality system based on a PDA with a compact flash (CF) multifunction I/O board

The sensing node includes sensors (temperature, relative humidity, and air quality), conditioning circuits, a compact flash data acquisition device DAQ (NI CF-6004) and a PDA with wireless communication capabilities (Wi-Fi or Bluetooth). A point-to-point connection between the measurement node and an advanced processing and communication unit (a PC) permits to deliver the air quality data from the sensing node to the PC and to receive information, such as alarm thresholds, that is used to implement alarm mechanisms in the PDA. The acquired data is processed by the PDA and the results are displayed by the PDA GUI.

Considering the cost of the implementation of the air quality sensing node based on a DAQ board plugged to a PDA, and also taking into account the evolution of the area of pervasive computing,, the authors decided to develop air quality monitoring systems based on smart phones and Bluetooth enabled smart sensors. The implemented architecture is presented in





#### Figure 10.

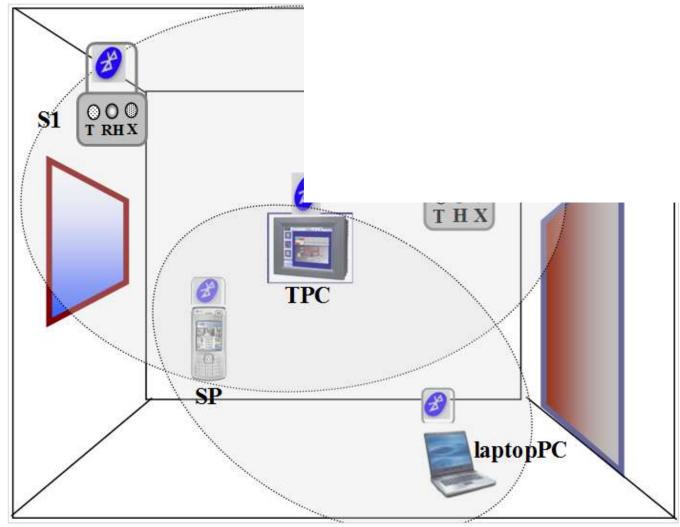
Air quality virtual measuring system's architecture based on a smart sensing node (SN) with Bluetooth communication capabilities, and on a smart phone

are specialized for temperature, air quality and air quality index measurement [ ]. When monitoring large spaces, the The smart sensors indicated in number of sensing nodes increases, which means that point-to-multipoint architectures must be considered.

B. Point-to-multipoint Bluetooth architecture and embedded smart phone software An implementation of a point-to-multipoint network architecture that uses Bluetooth compatible smart sensing nodes is presented in provide information about the level of relative humidity, temperature, and air contaminants (e.g. undesired odours that can trigger respiratory disorders). As computation units and human machine interface are included a laptop PC that works as the system server, a touch panel computer (TPC) and a smart phone

The implemented Bluetooth scatter net architecture assures the remote monitoring of the sensing nodes and data communication between the mobile device and smart sensor nodes. The hardware component of the system includes: sensors and conditioning circuits, a data acquisition device Bluetooth enabled (e.g. BlueSentry from Grid Connect), a smart phone with Bluetooth interface (e.g. N70 from Nokia), a situated display (NI TPC2106) Bluetooth compatible through a RS232-to-Bluetooth bridge, and a data communication, data processing and data storage unit (laptop PC).





#### Figure 11.

Distributed air quality measurement architecture associated with respiratory distress triggering factors monitoring based on Bluetooth networking protocol (S1 and S2 are the sensing nodes characterized by T-temperature, H - relative humidity and X- air quality index measurement channel, TPC- touch panel computer, SP- smart phone)

The software technologies used to develop the applications for the smart phone running Symbian OS and for the TPC running Windows CE OS, were Java2Me and LabVIEW. The application embedded in the smart phone was named SmartSense Mobile. AirQUbicomp is the application developed using LabVIEW 8.6 Touch Panel Module for the TPC. This application provides the information about indoor air quality.

The SmartSense application has the ability to identify the active smart sensing nodes, to establish a connection via Bluetooth with the nodes, to control the on/off state of the air quality index sensor (XairQ-sensor), and to collect voltage samples from relative humidity, temperature and air quality measuring channels of each node in single-shot mode or in continuous mode.

SmartSense also assures the transfer of the indoor air quality values calculated and stored in the smart phone memory extension to the laptop PC through Bluetooth synchronization.

After node(s) selection, the operator can choose between the "one sam; in order to test the normal functioning of the sensing node or to verify t broadband pollution) during the system setup.



Working in continuous acquisition mode, the smart phone application p thresholds previously stored in the SmartSense Mobile configuration fil files received through Bluetooth from the PC that runs a SmartAdmin a to 60 min interval were considered. These values are adapted to the small temperature, humidity and XairQ index measurement. During continuo converted into physical values by the SmartSense Mobile application an phone whose number was written in the SmartSense configuration file.

The continuous acquisition and data conversion software modules work inform that indoor air conditions are critical. The used threshold values

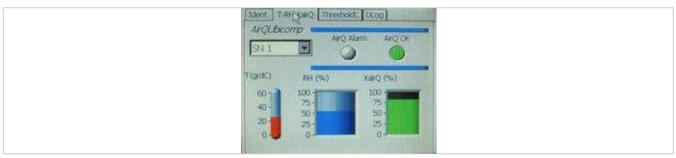
Th	RH[%]
thmin	30
thmax	50

#### Table 1.

Threshold values for relative humidity, temp likelihood of asthma attack.

During visual or acoustic signalling, a set of useful recommendations related to indoor air factors values and the actions necessary to change the indoor air conditions from critical to normal are available through the smart phone GUI.

The AirQUbicomp application is designed to continuously monitor the air quality, generating visual and acoustic alarms according to the imposed thresholds. Active interaction with the touch panel computer is permitted after identification of the user through a numeric password. After identification, the user can modify the thresholds related to asthma or can define data logging elements such as the time between readings and the monitoring period (DLog TAB in The values of temperature, air quality and air quality index as well as the alarms LEDs (AirQ Alarm) are part of the T-RH-XAirQ software TAB. In GUI associated with AirQUbicomp is presented.



#### Figure 12.

#### AirQUbicomp GUI

Using the developed SmartSense Mobile application different tests associated with indoor air quality monitoring were carried out. The data stored in the Nokia N70, is wireless transferred to the database implemented in the laptop PC. Some data related with continuous measurement of the asthma or COPD attack triggering factors are presented in

the relative humidity is in the limit of automatic alarm generation (RH>50%) while temperature and air quality are inside the interval values associated with "no asthma or COPD attack conditions".

can be observed low levels of the XairQ index when the measurement session started. Based on the information displayed, the user acted to improve the air quality (e.g. by opening the window). The air quality started to improve and, at the same time, room's temperature and humidity change significantly.



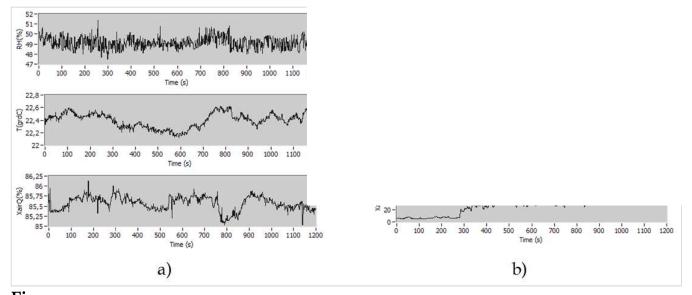


Figure 13. S1 node monitoring of respiratory distress triggering factors

In order to find correlations between the air quality and the values of physiological parameters, such as oxygen saturation (SpO2) and heart rate (HR), a digital pulse oxymeter and electrocardiograph apparatus ECG Medlab P-OX 100 was used for testing purposes. volunteers, with and without respiratory distress history (RD-N, RD-Y). The physiological values were measured in the same room and for the volunteers seated on a chair.

, one can notice that in case of the healthy individual (RD-N), values of XairQ lower than 80% and of RH near 50% do not induce Analyzing the data from changes in HR and SpO2, while a significant increase in the HR of RD-Y is felt.

Sensor T node (°C)	_	RH (%)	XairQ(%)	RD-N		RD-Y	
	H11 (70)	71an Q(70)	HR	SpO2	HR	SpO2	
S1	17.2	47.4	62.7	72	98	96	92
S2	17.8	44.2	69.6				

#### Table 2.

S1 node: air quality and physiological parameter values for two volunteers, with and without respiratory distress history

Nowadays, smart phones are provided with operating systems, such as Android OS and iOS, which make the implementation of complex software modules easier and faster. The authors have been working to develop an AirQ Android OS application for a multichannel sensing node. The graphical interface of the implemented application is presented in

The AirQ dashboard includes elements related with respiration activity (respiration rate). The data logging procedure is done using a smart phone embedded database that can synchronize with Web-based information system database through Wi-Fi or 3G/UMTS communication protocol.





Figure 14.

AirQ graphical interface implemented in the AndroidOS smart phone

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5. Conclusion
The quality of life of pulmonary patients greatly depends on the quality of the air they breathe. The identification of the indoor air associated with pathophysiology of COPD and asthma disease is crucial for the primary-prevention strategy. In the preceding paragraphs the authors summarize the main elements of a distributed smart sensing network for indoor air quality assessment. Regarding sensing nodes and signal conditioning, two possible solutions were presented. One based on semiconductor heated sensors and another based on three electrodes' cells. For data processing purposes, a hybrid solution based on relements and artificial neural networks modelling is presented. The last part of the chapter includes possible solutions for indoor air quality smart sensing polynomial and artificial neural networks modelling is presented. The last part of the chapter includes possible solutions for indoor air quality smart sensing networks.

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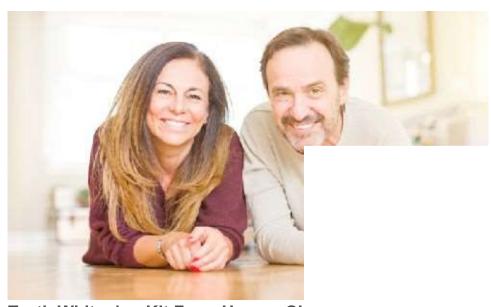




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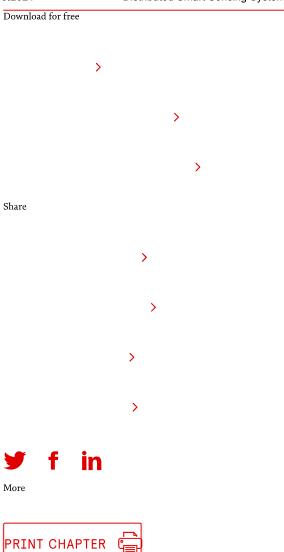


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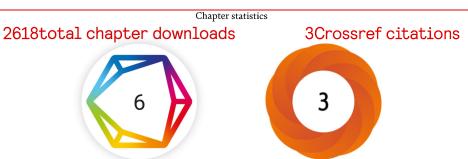
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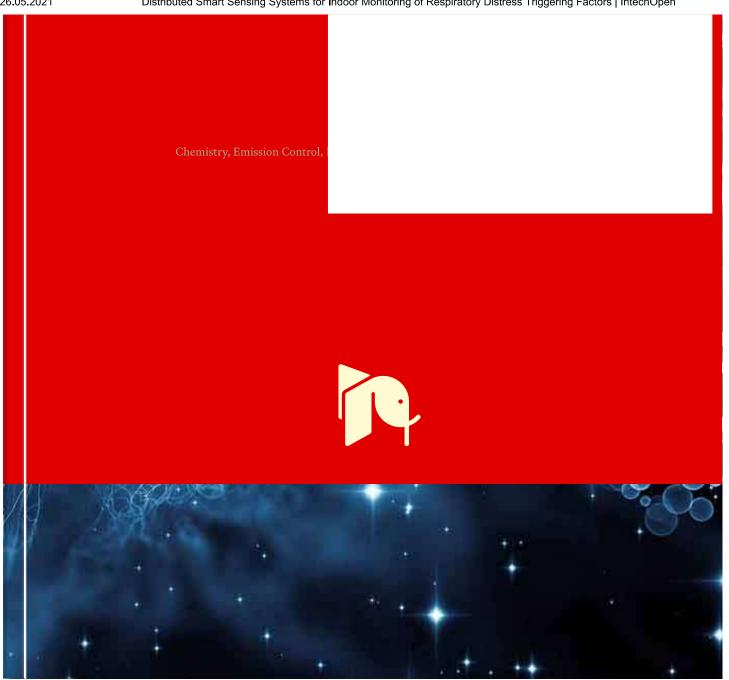
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